

Lecture-5: Multi- Antenna OFDM Systems

- Diversity Techniques in **OFDM**
- Receive Diversity in OFDM **Systems**
- **3. Transmit Diversity in OFDM Systems**
- 4. Spatial Multiplexing (MIMO-OFDM)
- 5. Smart Antenna Technique: **Beamforming**

Transmit Diversity

Time diversity \rightarrow Delay diversity S-T Diversity → STBC, STTC S-F Diversity → SFBC, SFTC S-T-F Diversity → STFBC

7. Spatial diversity coding for MIMO-OFDM

- \bullet • In MIMO system the Alamouti scheme realizes full spatial diversity gain in the absence of channel knowledge at the transmitter.
- \bullet This requires that the channel remains constant over at least two consecutive symbol periods
- \bullet • In MIMO-OFDM the coding is performed in the frequency rather than in time
- • Symbols s1 and s2 are transmitted over antennas 1 and 2 on tone ⁿ and symbols –s2* and s1* are transmitted over antennas 1 and 2 on tone n+1 OFDM symbol 1

7. Spatial diversity coding for MIMO-OFDM

- \bullet Any pair of tones could be used as long as the associated channels are equal i.e. the channel requirement is different from the MIMO case
- \bullet An alternative technique is to use diversity coding on ^a per-tone basis across OFDM symbols in time but then the channel should be constant during two consecutive OFDM symbols
- \bullet This is not usually true due to the long duration of OFDM symbols

Space-Time Block Code - II

- The received signals are $y_1 = \alpha_1 x_1 + \alpha_2 x_2 + \eta_1$ $*$ * $2 - \omega_1 \omega_2 + \omega_2 \omega_1 + \gamma_2$ $y_2 = -\alpha_1 x_2 + \alpha_2 x_1 + \eta_2$
- Calculate the decision variables as

$$
\hat{x}_1 = \alpha_1^* y_1 + \alpha_2 y_2^* = (|\alpha_1|^2 + |\alpha_2|^2) x_1 + \alpha_1^* \eta_1 + \alpha_2 \eta_2^*
$$

$$
\hat{x}_2 = \alpha_2^* y_1 - \alpha_1 y_2^* = \left(\left| \alpha_1 \right|^2 + \left| \alpha_2 \right|^2 \right) x_2 - \alpha_1 \eta_2^* + \alpha_2^* \eta_1
$$

• Similar to that of a two-branch maximal ratio combining receiver diversity system!

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Space-Time Block-Coded OFDM - I

• Space-time coding on two adjacent blocks of data symbols, i.e., $\mathbf{X}(n)$ and $\mathbf{X}(n+1)$.

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STBC-OFDM Simulation Results

- STBC-OFDM achieves near optimal diversity gain in slow fading.
- Still outperforms non-diversity OFDM system at $f_D = 100$ Hz.

Space-Frequency Block-Coded OFDM - I

• Coding on adjacent DFT frequency bins of each block of **X**(*n*).

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Space-Frequency Block-Coded OFDM - II

• Space-frequency encoder codes each data vector **X**(*n*),

$$
\mathbf{X}(n) = \left[X(n,0) \ \ X(n,1) \cdots X(n,K-2) \ \ X(n,K-1) \right]^T,
$$

into two vectors $\mathbf{X}_1(n)$ and $\mathbf{X}_2(n)$ as

$$
\mathbf{X}_{1}(n) = \begin{bmatrix} X(n,0) - X^{*}(n,1) \cdots X(n,K-2) - X^{*}(n,K-1) \end{bmatrix}^{T},
$$

\n
$$
\mathbf{X}_{2}(n) = \begin{bmatrix} X(n,1) & X^{*}(n,0) \cdots X(n,K-1) & X^{*}(n,K-2) \end{bmatrix}^{T},
$$

or in terms of the even and odd polyphase vectors as $\mathbf{X}_{_{1,e}}\left(n \right)=\mathbf{X}_{_{e}}\left(n \right) \quad , \quad \mathbf{X}_{_{1,o}}\left(n \right)=-\mathbf{X}_{_{o}}^{^{*}}\left(n \right)$ $\mathbf{X}_{_{2,e}}(n)$ $=$ $\mathbf{X}_{_{o}}(n)$ \quad , $\quad \mathbf{X}_{_{2,o}}(n)$ $=$ $\mathbf{X}_{_{e}}^{^{*}}(n)$ \mathbf{A}_{e} (*n*) = \mathbf{A}_{o} (*n*) = $\mathbf{A}_{2,o}$ (*n*) = \mathbf{A}_{e} (*n*) $X_1(n) = X(n)$, $X_2(n) = -X^*$ $X_{2}(n) = X(n)$, $X_{3}(n) = X^{*}$

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Space-Frequency Block-Coded OFDM - III

- The demodulated vector is or, equivalently, as ${\bf Y}(n) = {\bf \Lambda}_{1}(n) {\bf X}_{1}(n) + {\bf \Lambda}_{2}(n) {\bf X}_{2}(n) + {\bf Z}(n),$ $\mathbf{Y}_{e} \left(n \right) = \boldsymbol{\Lambda}_{1,e} \left(n \right) \mathbf{X}_{1,e} \left(n \right) + \boldsymbol{\Lambda}_{2,e} \left(n \right) \mathbf{X}_{2,e} \left(n \right) + \mathbf{Z}_{e} \left(n \right)$ $\mathbf{M}_{1, o} (n) = \mathbf{\Lambda}_{1, o} (n) \mathbf{X}_{1, o} (n) + \mathbf{\Lambda}_{2, o} (n) \mathbf{X}_{2, o} (n) + \mathbf{Z}_{o} (n)$. $\mathbf{Y}_{o} (n) = \mathbf{\Lambda}_{1,o} (n) \mathbf{X}_{1,o} (n) + \mathbf{\Lambda}_{2,o} (n) \mathbf{X}_{2,o} (n) + \mathbf{Z}_{o} (n)$
- Calculate $(n) = \Lambda_{1,e}^{*}(n)Y_{e}(n) + \Lambda_{2,o}(n)Y_{o}^{*}(n)$ $\mathbf{M}(n) = \mathbf{\Lambda}_{2,e}^{*}(n) \mathbf{Y}_{e}(n) - \mathbf{\Lambda}_{1,o}(n) \mathbf{Y}_{o}^{*}(n)$ $1,e \vee e$ \vee $e \vee e$ \vee e \vee e 2, 1, ˆ \mathbf{A}_{e} (*n*) = $\mathbf{\Lambda}_{1,e}$ (*n*) \mathbf{Y}_{e} (*n*) + $\mathbf{\Lambda}_{2,o}$ (*n*) \mathbf{Y}_{o} (*n*) . ˆ \mathbf{A}_{o} (*n*) = $\mathbf{\Lambda}_{2,e}$ (*n*) \mathbf{Y}_{e} (*n*) – $\mathbf{\Lambda}_{1,o}$ (*n*) \mathbf{Y}_{o} (*n*) $\mathbf{X} (n) = \mathbf{\Lambda}^* (n) \mathbf{Y} (n) + \mathbf{\Lambda} (n) \mathbf{Y}^*$ $\mathbf{X} (n) = \mathbf{\Lambda}_{\circ}^{*} (n) \mathbf{Y} (n) - \mathbf{\Lambda}_{\circ} (n) \mathbf{Y}^{*}$
- Assuming $\Lambda_{1,e}(n) \approx \Lambda_{1,o}(n)$ and $\Lambda_{2,e}(n) \approx \Lambda_{2,o}(n)$, yields $\left({{{\left| {{\bf{\Lambda }}_{1,e}} \right|}^2} + {{\left| {{\bf{\Lambda }}_{2,e}} \right|}^2}} \right)$ $\left({{{\left| {{\mathbf{\Lambda }}_{1,o}} \right|}^2} + {{\left| {{\mathbf{\Lambda }}_{2,o}} \right|}^2}} \right)$ $1,e$ \uparrow \uparrow $2,e$ \uparrow \uparrow \uparrow e \uparrow \uparrow $1,e$ \downarrow e \uparrow \uparrow $2,$ ˆ $\mathbf{A}_{e} = \begin{bmatrix} 1 & 1 \\ 1 & 2 \end{bmatrix}$ $\mathbf{A}_{2,e} = \begin{bmatrix} 1 & 1 \\ 1 & 2 \end{bmatrix}$ $\mathbf{A}_{e} = \begin{bmatrix} 1 & 1 \\ 1 & 2 \end{bmatrix}$ ˆ $\mathbf{X} = \left(\begin{matrix} \mathbf{A} & \mathbf{A} \end{matrix} \right) + \left(\mathbf{A} \right) \left(\begin{matrix} \mathbf{A} & \mathbf{B} \end{matrix} \right) + \mathbf{A} \left(\begin{matrix} \mathbf{X} & \mathbf{A} \end{matrix} \right) + \mathbf{A} \left(\begin{matrix} \mathbf{Z} & \mathbf{A} \end{matrix} \right)$ $\mathbf{X} = \left(\begin{matrix} \mathbf{A} & \mathbf{A} & \mathbf{A} \end{matrix} \right) + \left(\mathbf{A} \mathbf{A} \right) \left(\begin{matrix} \mathbf{X} & \mathbf{A} & \mathbf{X} \\ \mathbf{X} & \mathbf{A} & \mathbf{Z} \end{matrix} \right) + \mathbf{A} \left(\begin{matrix} \mathbf{X} & \mathbf{A} & \mathbf{X} \\ \mathbf{X} & \mathbf{A} & \mathbf{Z} \end{matrix} \right)$

 $1, o \quad$ \uparrow \uparrow $2, o \quad$ \uparrow \uparrow $o \quad$ \uparrow \uparrow $2, e \rightarrow$ $e \quad$ \uparrow \uparrow \uparrow

 $\mathbf{A}_{o} = \begin{bmatrix} 1 & 1 & 0 \\ 0 & 0 & 0 \end{bmatrix}$ $\mathbf{A}_{2,o} = \begin{bmatrix} 1 & 1 & 0 \\ 0 & 0 & 0 \end{bmatrix}$

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SFBC-OFDM Simulation Results - I

- SFBC-OFDM achieves similar diversity gain as STBC-OFDM in slow fading.
- SFBC-OFDM performs better in fast fading.

SFBC-OFDM Simulation Results - II

- STBC-OFDM is more sensitive to channel gain variation over time.
- SFBC-OFDM is more sensitive to channel gain variation over frequency.

ST(F)BC Decoding ST(F)BC Decoding

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What is a MIMO System??

■ What are MIMO systems ?

A MIMO system consists of several antenna elements, plus adaptive signal processing, at both transmitter and receiver, the combination of which exploits the spatial dimension of the mobile radio channel.

Benefits

- higher capacity (bits/s/Hz):
	- spectrum is expensive; number of BS is limited
- better transmission quality
- Increased coverage
- Improved user position estimation

Narrowband MIMO Channel Model

• Frequency flat channel model

Wideband MIMO Channel Model Wideband MIMO Channel Model

 Channel measurements typically show large delay spreads. **Insufficient antenna spacing or lack of scattering cause** individual antennas to be correlated.

Geometry of Ricean component has a direct bearing on multiplexing gain.

Ξ Broadband channel is frequency selective :

$$
H(\tau) = \begin{bmatrix} h_{1,1}(\tau) & h_{1,2}(\tau) & \dots & h_{1,N_T}(\tau) \\ h_{2,1}(\tau) & h_{2,2}(\tau) & \dots & h_{2,N_T}(\tau) \\ M & M & M & M \\ h_{N_R,1}(\tau) & h_{N_R,2}(\tau) & \dots & h_{N_R,N_T}(\tau) \end{bmatrix}
$$

MIMO System Model MIMO System Model

with

Noise: $n_{\scriptscriptstyle \hat{i}}$: $\mathbf{CN}\left(0,\sigma_{\!R}^2\right)$ with $\mathrm{E}\!\left\{m^H\right\}\!=\!\sigma_{\!R}^2 I_{M_{\!R}}$ r_1 r_2 \ldots r_{N_R} | \ldots N_R \times 1 receive signal vector $\begin{array}{cc} \mathbf{x}_1 & \mathbf{x}_2 & ... & \mathbf{x}_{N_T} \end{array}$ | $\begin{array}{cc} ... & N_T \times 1 \end{array}$ transmit signal vector η_1 n_2 ... $n_{\scriptscriptstyle N_R}$... $\text{N}_{\scriptscriptstyle R} \times 1$ noise vector $\boldsymbol{\mathrm{H}}$... $\boldsymbol{\mathrm{N}}_{\!R}\!\times\!\!\boldsymbol{\mathrm{N}}_{\!T}$ channel transfer matrix *T* $r = |r_1 \r_2 \dots r_{N_p}| \dots N_R \times$ *T* $x = \begin{vmatrix} x_1 & x_2 & \dots & x_{N_r} \end{vmatrix}$ \ldots $N_T \times$ *T* $n = \begin{vmatrix} n_1 & n_2 & ... & n_{N_R} \end{vmatrix}$ $... N_R \times$ $\begin{bmatrix} r_1 & r_2 & ... & r_{N_R} \end{bmatrix}^T$... N_R $\begin{bmatrix} x_1 & x_2 & \dots & x_{N_T} \end{bmatrix}^T$ $\ldots N_T$ $\begin{bmatrix} n_1 & n_2 & ... & n_{N_R} \end{bmatrix}^T$... N_R

Lecture-6 Outline:

Multi-Carrier Based Multiple Access Schemes

- **1.Introduction**
- 2.Basics of OFDMA
- 3. OFDMA in Uplink
- 4.OFDM-CDMA
- 5. Discussion about Access Techniques

Introduction

- \Box In recent years, OFDM has gained prominence in high data rate WLAN
- \Box Though OFDM provides efficient transmission scheme at high data rate, it does not possess any inherent multi-user capability
- \Box The objective of this lecture is to understand the methods to instantiate multiple access communications via the multi-carrier approach
- \Box The concentration will be on cellular systems based on multi-carrier techniques, thus access techniques for downlink (DL) and uplink (UL) will be presented.

Basic Multiple Access Techniques

- \Box Multiple access schemes are used to allow many users to share simultaneously a finite amount of spectrum.
- \Box The sharing of spectrum is required to achieve high capacity by simultaenously allocating the available bandwidth to multiple users.
- \Box For high quality communicatios, this must be done without severe degradation in the performance of the system.
- \Box There are many access techniques , some of them are
	- \checkmark Frequency Division Multiple Access (FDMA)
	- \checkmark Time Division Multiple Access (TDMA)
	- \checkmark Space Division Multiple Access (SDMA)
	- Spread Spectrum Multiple Access (SSMA)
		- \checkmark Code Division Multiple Access (CDMA)
		- \checkmark Frequency Hope Multiple Access (FHMA)

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Generic OFDM Based Multiple Access Model (1) Generic OFDM Based Multiple Access Model (1)

Downlink and Uplink Scenario Downlink and Uplink Scenario

■ One transmitter and multiple receiver in DL **DL** is similar to broadcast systems

- Several transmitters and one receivers in UL. Thus, UL is an ideal multi-access channel.
- All transmitters have unique time and frequency offset, thus, UL system design is more difficult than DL.
- **For future systems, asymmetric traffic distribution** between UL and DL is expected. Thus, the access technique should be able to support high data rate in DL and a moderate data rate in UL.

Available Techniques

For Seamless Area Coverage, Multi-cell Structure is necessary.

CDMA achieves the best spectrum efficiency.

For High Throughput, Single-cell Structure is suitable because it avoids inter-cell interference.

OFDM achieves the best spectrum efficiency.

However

- Each scheme is not best in the other cell structure.
- For the flexible area coverage and service deployment with lower cost, both Multi-cell and Single-cell environments should be supported with maximum throughput.

Source: NTT DoCoMo 2002

Need for **New Wireless Access Scheme**that covers **both environments.**

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OFDMA

A fraction of OFDM sub-carriers is assigned to a user in a contiguous or interleaved manner Blocked OFDMA: simple but possible throughput degradation due to channel fading Interleaved OFDMA: attains channel diversity but there is need for 'stricter' synchronization

OFDMA Transceiver Architecture OFDMA Transceiver Architecture

 User data is modulated with a baseband modulated scheme**□ Baseband modulated** symbols are assigned to sub-carriers, using the assignment map defined by the subcarrier allocation scheme, then OFDMA symbol is transmitted.

 \square Data of u^{th} user can be received by knowledge of the subcarrier assignment

OFDMA: Specific Features OFDMA: Specific Features

\Box Very simple scheme to implement

- \Box In downlink OFDMA has all the advantages and disadvantages of basic OFDM, with the addition of multiple access capability. Thus, a high peak data rate can be achieved in DL through OFDMA.
- \Box In uplink time and frequency synchronization becomes important (We will discuss this later in details).
- □ OFDMA with high user mobility: When some of the users are moving with a high speed, then the coherence time is small and Doppler effect is severe, then special measure needs to be taken due to the relation between channel coherence time and OFDM symbol duration

Static and Dynamic Sub-carrier Allocation

Each user can have a pre-determined set of sub-carriers for the duration of connection in Static Sub-carrier Allocation (SSA). Alternatively, number of sub-carriers can be varied based on the current traffic, user the channel and the required user data rate to transmit in Dynamic Sub-carrier Allocation (DSA).

- $\mathsf{SSA}:\ \ \circledcirc\ \mathsf{Simple}\ \mathsf{to}\ \mathsf{implement}$
	- ☺ Minimal signaling overhead
	- No frequency diversity is achieved
	- \odot Sub-optimal throughput, at times may be below the expected rate
- DSA: © Optimized throughput
	- Channel State Information (CSI) of all the users for all sub-carriers are required in real-time.
	- \odot Signaling overhead is increased compared to SSA.

Dynamic Allocation Algorithm Dynamic Allocation Algorithm

□ For sub-carrier allocation specific algorithms are required at the base station to assign the sub-carriers.

□ The objective functions of the allocation algorithms can be either thorough optimization or QoS or both.

□ Some of the well-known algorithms are:

- **→ Round Robin Scheduling**
- → Max C/I based Scheduling
- → Proportional Fair Scheduling

Round Robin and Max C/I in Brief Round Robin and Max C/I in Brief

Practicality of Dynamic Sub-carrier Allocation

□ DSA can be impractical especially when

- Coherence time is very low, that the channel time variability is high
- Large number of sub-carriers are present in DFT window
- Large number of users require simultaneous services in the system

This is because

- excessive signaling is required when CSI related to large number of sub-carriers and large number of users are transmitted to BS
- \triangleright the required computation time of scheduling algorithm will be very high.

□ One alternative to DSA is sub-carrier hopping.

Sub-Carrier Hopped OFDMA (SCH-OFDMA)

 Users hop between sub-carriers from OFDMA symbol to symbol in SCH-OFDMA.

□ When users hop after every OFDMA symbol, then we call it Fast SCH-OFDMA (FSCH-OFDMA). Similarly, when users hop after some OFDMA symbols, then we call it Slow SCH-OFDMA.

Using SCH-OFDMA

- Minimizes the inter-cell interference
- Averages the impact of fading among users
- Does not require CSI for sub-carrier allocation anymore.
- Near Far resistance
- □ Definitely, the throughput is reduced compared to dynamic allocation with the knowledge of CSI, but greatly enhanced compared to static allocation

Time- Frequency Diagram of FSCH Frequency Diagram of FSCH-OFDMA

Transmitter Side: From the Base Station

 \rm{MS} 2

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 $T_{\rm s}$

 $\overline{0}$

 $[\n\Gamma_{\rm c}/T_{\rm s}]T_{\rm s}$

2 $\left[\begin{array}{c} C \\ C \end{array} \right]$ T_s

Time

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SCH-OFDMA: Specific Features

- \Box Orthogonal hopping sequences used among users \rightarrow no Intra-cell interference
- □ Each BS has an orthogonal hop-set compared to neighboring cells
	- \rightarrow Inter-cell interference can be prevented
- □ Orthogonal sub-carriers and perfect synchronization \rightarrow no ICI within each cell
- \Box If CP larger than maximum excess delay \rightarrow no ISI
- \Box If T_s < T_c, then sub-carrier indices allocated to each user change faster than the channel, then the 'bad' sub-carriers are distributed among users
	- \rightarrow error rate of each user is reduced

Choosing Hopping Rate Choosing Hopping Rate

Since in outdoor coherence time is smaller than in indoor, SSCH tends to coincide with FSCH

Receiver has to hop synchronously with the transmitter to recover the transmitted information

In DL it is easier to synchronize compared with UL, since in the DL transmissions experience a single common channel to reach a user, while in UL there are several channels from users to the BS

SSCH has lower signaling complexity for synchronization than FSCH, since hops occur less frequently

SCH-OFDMA Based System: Flash OFDM

- \Box "Fast Low-latency Access with Seamless Handoff (FLASH) OFDM'' proposed by Flarion Inc.
- \Box Flash-OFDM uses FSCH in the downlink and SSCH in the uplink.
- \Box Approved in IEEE 802.16a WMAN and under consideration in IEEE 802.20 system.

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Conventional System Model for Uplink OFDMA

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 Different types of wireless channel, i.e. different coherence time and coherence bandwidth is expected from different user □ The physical distance of a user can be very large in large cells

OFDMA Uplink Scenario OFDMA Uplink Scenario

Latures Users are transmitting from distant locations to BS with

- different timing offset (DFT window de-synchronization)
- different carrier frequency offset (oscillator mismatch)
- different Doppler frequency (due to user motion)
- Carrier frequency offset = MS oscillator offset + BS oscillator mismatch and total frequency offset = carrier frequency offset + Doppler spread
- я, Timing offset = f(maximum delay spread, propagation path delay)
- я, These offsets are unique to all users, and each user's offset needs to be resolved separately.

Timimg Synchronization Timimg Synchronization

- \Box All incoming signals must be aligned with BS's FFT window, otherwise both ICI and ISI are introduced
- \Box The relatively different delays between incoming signals reduce OFDM's robustness to delay spread
- \Box Multiuser timing synchronization is not always possible

Some Known Solutions for Uplink OFDMA Issues

Increasing the CP interval to account for differential delays: \rightarrow Induce SNR loss and reduce system throughput

Timing and frequency synchronization is done at MS:

- □ BS performs estimation and sends feedback information to MS,
- □ MS adjusts its timing and frequency base to account for the mis-matches,
- \rightarrow Require established connection between MS and BS, which is not applicable in some scenarios (e.g. random access)

Multi-user detection and separation before individual time- and frequency-estimation and correction

- □ Work for both contiguous and non-contiguous subcarrier allocation schemes
- \rightarrow However, this method require much more processing power at BS for multi-user detection, separation and decoding

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OFDM-CDMA: Main Idea (1/2)

- \Box OFDM overcomes ISI effect of multi-path frequency selective wireless channel, while CDMA provides frequency diversity and multi-user access scheme
- \Box The idea of sharing time, frequency and energy between different users can be depicted in the figures

(a) Frequency-domain spreading; (b) Time-domain spreading; (c) Time-Frequency localized spreading, all with SF=8

OFDM-CDMA: Main Idea (2/2)

- \Box Only time and frequency domain in the figure. Each data symbol is spread either in frequency domain, or time domain or in both. Coding plane is not shown for clarity.
- \Box In figure, spread chips of a symbol occupy contiguous subcarriers. In practice, chips of different symbols can be interleaved to achieve frequency diversity.
- \Box At the very first phase in 1993, three slightly different schemes were independently proposed:
	- □ MC-CDMA (Yee, Linnartz, Fettweis, and others)
	- Multi-Carrier DS-CDMA (DaSilva and Sousa)
	- Munti-Tone CDMA (Vandendorpe)
- \Box Number of symbols to be transmitted in one OFDM symbol is an implementation issue, related to traffic load, system requirement and transmission scenario.

MC-CDMA (Multi-Carrier CDMA)

- MC-CDMA implements frequency domain spreading via orthogonal codes, thus, several users transmit over the same subcarrier. The transmitter spreads the data, then modulates a different subcarrier with each chip, thus spreading in frequency domain is realized.
- MC-CDMA has been so far hailed to be a very efficient in DL. Asynchronous UL has a severe effect on orthogonal codes
- \Box If c(t)=[c₁, c₂, ..., c_N] represents the spreading code for an arbitrary user, the transmitter structure is:

DS-CDMA and MC CDMA and MC-CDMA

MC-CDMA has similar structure as DS-CDMA except:

- □ Spreading codes is applied in frequency domain.
- □ Signal is transmitted over several sub-carriers that constitute the entire spectrum. Therefore, same amount of bandwidth is used for both the system

Since spreading codes is applied in frequency domain, information bits are not chopped into smaller chips.

Therefore, the each bit is longer than the chip in DS-CDMA system. So, MC-CDMA is less sensitive to the channel coherence bandwidth.

Downlink MC-CDMA Transmitter

Downlink MC-CDMA Receiver

MC-CDMA: Specific Features

- \Box It is one of the candidates that can be used as one of the future DL access schemes, since it provides high frequency diversity, low complex equalization and high spectral efficiency
- \Box Variation of the scheme: OFDMA-CDM \rightarrow one user is given a set of sub-carriers. All the data symbols that the user has to transmit are spread and transmitted over those sub-carriers only.
- □ OFDMA-CDM does not introduce MAI as MC-CDMA does, but it introduces self-interference, it can be a viable scheme for UL in small indoor scenarios
- □ OFDMA-CDM with Frequency Hopping avoids near far effect or asynchronous transmission (might be OK for UL)

Time Domain Spreading and Multi-carrier Modulation

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 Two schemes that combines time domain spreading on multicarrier modulation are:

- Multi-Carrier DS-CDMA Scheme
- Multi-Tone CDMA Scheme

■ Multi-Carrier DS-CDMA transmitter spreads the serial-toparallel(S/P)-converted data streams using a given spreading code in the time domain so that the resulting spectrum of each subcarrier satisfies the orthogonality condition with the minimum frequency separation. This scheme can lower the data rate in each subcarrier so that a large chip time makes it easier to synchronize the spreading sequences.

□ The MC-DS-CDMA scheme is originally proposed for a uplink communication channel, because this characteristic is effective for the establishment of a quasi-synchronous channel.

Multicarrier DS Multicarrier DS-CDMA

Beceived

 (C) Receiver

 $cos(2\pi f \text{M}Dt)$

 $\Gamma_{\rm m}$

Signal

LPF

LPF

- \Box Spreads the S/P converted data, then modulates a different subcarrier with each stream (spreading in time domain).
- \Box If c(t)=[c₁, c₂, ..., c_N] represents the spreading code for an arbitrary user, the transmitter structure is shown in the figures.

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Parallel

 $-10-$

Serial Converter

MT-CDMA scheme CDMA scheme

(a)Transmitter (b)Power spectrum of its transmitted signal (c)Receiver

MT-CDMA is very similar to MC-DS-CDMA, only that the orthogonality between the sub-carriers are lost in MT-CDMA, thus DFT demodulation cannot be used.

BEP Comparison in a Downlink Channel with 2-path i.i.d. Multipath Delay Profile (1/2) i.i.d. Multipath Delay Profile (1/2)

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Comments on Performance Comparison (1/2)

 MC-CDMA system has no major advantage over DS-CDMA system in terms of required bandwidth, because the bandwidth of MC-CDMA signal spectrum is almost the same as that of DS-CDMA signal spectrum.

□ In terms of transmission performance, the BEP lower bound of MC-CDMA system is all the same as that of DS-CDMA system. Therefore, If we make every effort to improve the BEP in each system, there is no difference in the attainable BEP as long as the same channel is used.

 DS-CDMA system cannot always employ all the received signal energy scattered in the time domain, whereas MC-CDMA system can effectively combine all the received signal energy scattered in the frequency domain, although it requires much complexity in the receiver structure such as subcarrier synchronization.

Comments on Performance Comparison (2/2)

 The MMSEC-based MC-CDMA must be a promising scheme in a downlink channel, although estimation of the noise power as well as subcarrier condition is required.

□ In the up-link application, a multi-user detection is required because the code orthogonality among users is totally distorted by the channel frequency selectivity.

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General Comments about the Schemes

- □ OFDMA with DSA is throughput optimal compared to all other scheme in DL. But implementing DSA may require excessive signaling. In UL, OFDMA requires special care because of time and frequency misalignment among multiple users.
- OFDM-TDMA is another simple scheme which can be implemented with less signaling compared to OFDMA. It can be a good solution for UL. Supporting real-time burst traffic is a problem in this technique.
- □ MC-CDMA performs very good in DL, but it suffers from MAI. So, OFDMA-CDM can be even better for DL. For UL, OFDM-CDMA based schemes are probably not a good solution due to loss of orthogonality and MAI.
- Compared to OFDMA and OFDM-TDMA, OFDM-CDMA is better in terms of frequency reuse factor. Reuse of factor of 1 Is possible in OFDM-CDMA, but at least 3 is required in OFDMA for a reasonable performance.

Outdoor Scenario: Outdoor Scenario: Channel Characteristics Channel Characteristics

- 1.Large cell radius
- 2.Large number of users
- 3. High mobility (high Doppler)
- 4.Large RMS delay spread
- 5. Wide distribution of user transmit power
- 6. Handover (re-synchronization of transceivers, signaling)
- 7.Inter-Cell interference

Outdoor Scenario: General Comment

 \Box If orthogonal codes are used while spreading the signals, inter-cell interference cannot be avoided. Choosing PN sequence instead can be an effective solution.

- □ Uplink: none of the schemes discussed so far can be considered the best possible. Thus there is a possibility of finding some new access schemes which perform better than the considered ones
- □ Downlink: inter-cell interference is the main concern
- Downlink: OFDMA with some form of DSA and/or OFDMA-CDM is a preferred choice, with its advantage of frequency diversity.
- □ Before suggesting any access, more detailed system level analysis with several impairments are required.

Indoor Scenario: Channel Characteristics

- 1.Small cell
- 2.Relatively small number of users
- 3. Very little handover problem
- 4. Low mobility, channel static for a long period so algorithms can be implemented for large number of symbols blocks
- 5. AMC and feedback systems can work in this environment
- 6. User transmit power uniformly distributed so little problem in power management
- 7.Low delay spread, and Large coherence bandwidth
- 8. Interference from Bluetooth networks, home networks, ISM band devices

Indoor Scenario: General Comment Indoor Scenario: General Comment

- \Box Inter-cell interference is not an issue, but the inter-network interference is still present for unlicensed technologies.
- \Box OFDMA is a simple scheme and since not many users are present, signaling is not very high in such a situation.
- \Box OFDMA and even OFDM-TDMA performance are expected to be quite good.
- \Box OFDM-CDMA is supposed to have the best quality, since it can exploit frequency diversity.
- \Box OFDMA-CDM is still a good option for DL.

Conclusion Conclusion

□ There are numerous other multiple access schemes that are not presented in this lecture.

 The goal is to give a high-level overview and overall understanding, thus, no mathematical expressions are presented. Interested students can refer to relevant references.

 Multiple access techniques for 4G is hot topic for research now.

□ Do contact us, if you are interested in the topics discussed in the lectures.