

ECE5984

**Orthogonal Frequency Division Multiplexing and Related
Technologies**
Fall 2007

Mohamed Essam Khedr

OFDM Basics I

Textbook

- **OFDM and MC-CDMA for broadband multi-user communications by Lajos Hanzo et al**
- **Additional Readings:**
 - **Richard van Nee and Ramjee Prasad, OFDM for Wireless Multimedia Communications, Artech House: 2000 (ISBN: OR90065306)**
 - **Orthogonal Frequency Division Multiplexing for Wireless Communications by Ye (Geoffrey) Li (Editor), Gordon L. Stuber (Editor), ISBN 0387290958**
 - **Ahmad Bahai and Burton Saltzberg, Multi-Carrier Digital Communications: Theory and Applications of OFDM, Plenum Publishing Corporation: 1999, ISBN: 0306462966.**

Syllabus

- **Wireless channels characteristics (7.5%)** **1**
 - wireless channel modeling and characteristics
 - Large scale and small scale models
 - Common channel models
 - Channel categories and parameter calculation.
 - **Prob. of error calculations**
- **OFDM Basics (10%)** **1**
 - History of OFDM
 - OFDM System model
 - Discrete-time signals & systems and DFT
 - Generation of subcarriers using the IFFT
 - Guard time, cyclic extension
 - Windowing
 - Choice of OFDM parameters & OFDM signal processing
 - Implementation complexity of OFDM versus single carrier modulation
- **Modulation and Coding (10%)** **2**
 - Linear and nonlinear modulation
 - Interleaving and channel coding
 - Optimal bit and power allocation
 - Adaptive modulation

Syllabus

- **Analysis of OFDM systems (15%)** **2**
 - RF subsystems, amplifier classification and distortion
 - Crest factor (PAPR) reduction techniques
 - Pre-distortion & adaptive pre-distortion techniques
 - clipping
 - coding techniques
 - partial transmit sequences (PTS) & modified PTS v. selective mapping
 - nonlinear quantization (companding)
 - Phase noise and I&Q imbalance for QAM
 - Performance of OFDM in Gaussian channels
 - Performance of OFDM in Wide-band channels
- **Synchronization and Estimation (15%)** **2**
 - ICI and OISI problems
 - Timing estimation
 - Frequency synchronization
 - Frequency error estimation algorithms
 - Carrier phase tracking
 - Frequency domain and time domain approaches for channel estimation
 - coherent detection
 - differential detection

Syllabus

- **Multi-user OFDM Techniques (10%)** 2
 - Adaptive modulations in OFDM
 - Power and bit allocations in OFDM
 - Scalable OFDM
 - Flash OFDM
- **Diversity (7.5%)** 1
 - Limits of capacity in fading environments
 - Channel models for multiple-input-multiple-output (MIMO) system
 - Receiver diversity techniques
 - Transmit diversity techniques and design criteria for fading channels
 - Block, trellis and layered space-time codes
- **Multi-carrier CDMA (10%)** 1
 - MC-CDMA versus DS-CDMA
 - MC-CDMA versus orthogonal frequency division multiple access (OFDMA)
 - OFDMA and MC-CDMA performance evaluation in wide-band channels

Syllabus

- **Physical and Medium Access Control (MAC) for IEEE 802.11 Networks**
(7.5%)
1
 - Physical modeling of 802.11 networks
 - MAC system architecture
 - Frame exchange with RTS/CTS
 - Power management
 - Synchronization
- **Physical and Medium Access Control (MAC) for IEEE 802.16 Networks**
(7.5%)
1
 - Physical modeling of 802.16 networks
 - MAC system architecture
 - QoS guarantees in Wimax
 - Power management
 - Synchronization

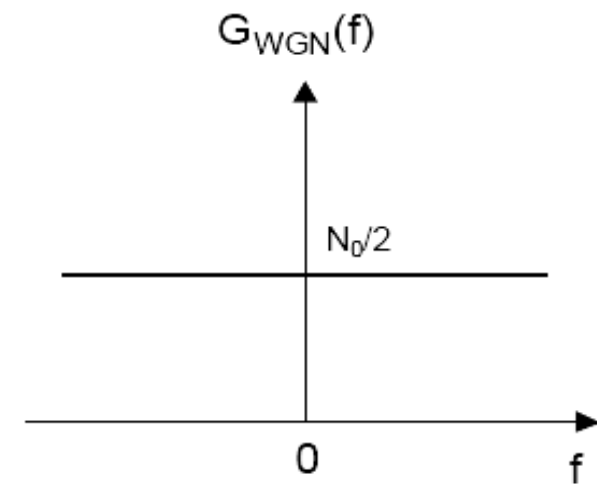
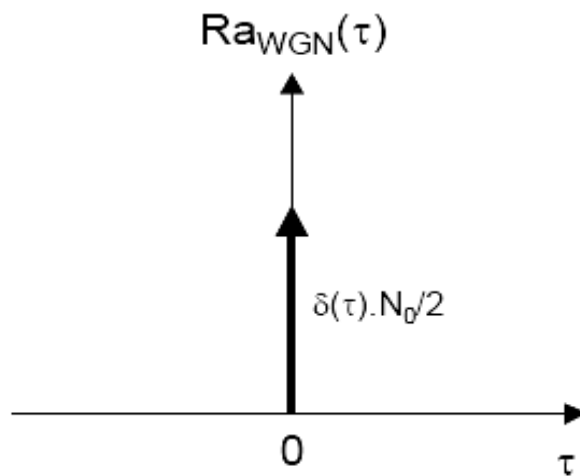
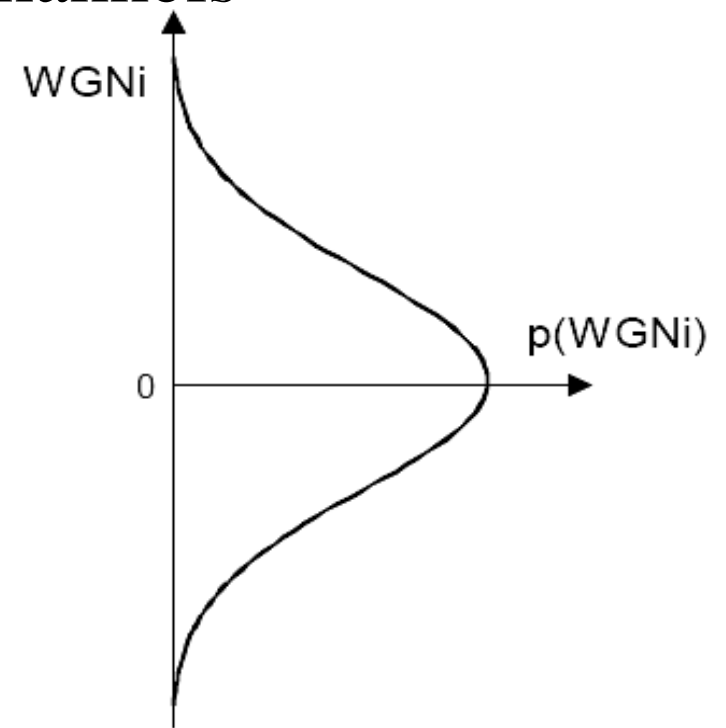
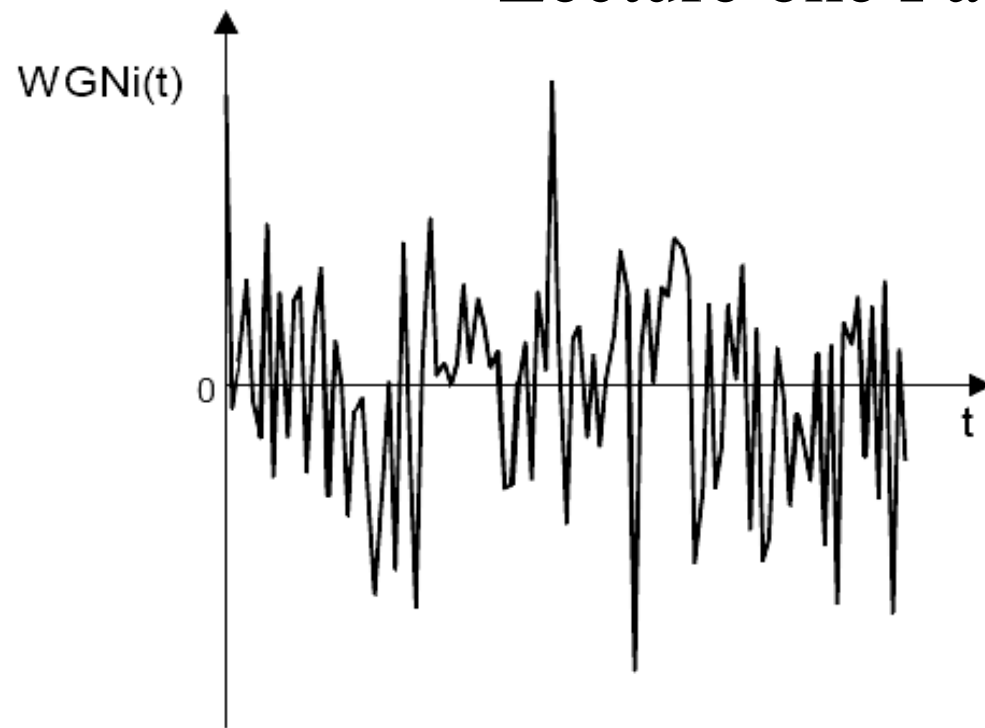
Grading

Type of assignment	Percent of Grade
Home works	20%
Matlab Assignments	20%
Midterm	20%
Final project presentation and term paper	20%
Final Exam	20%

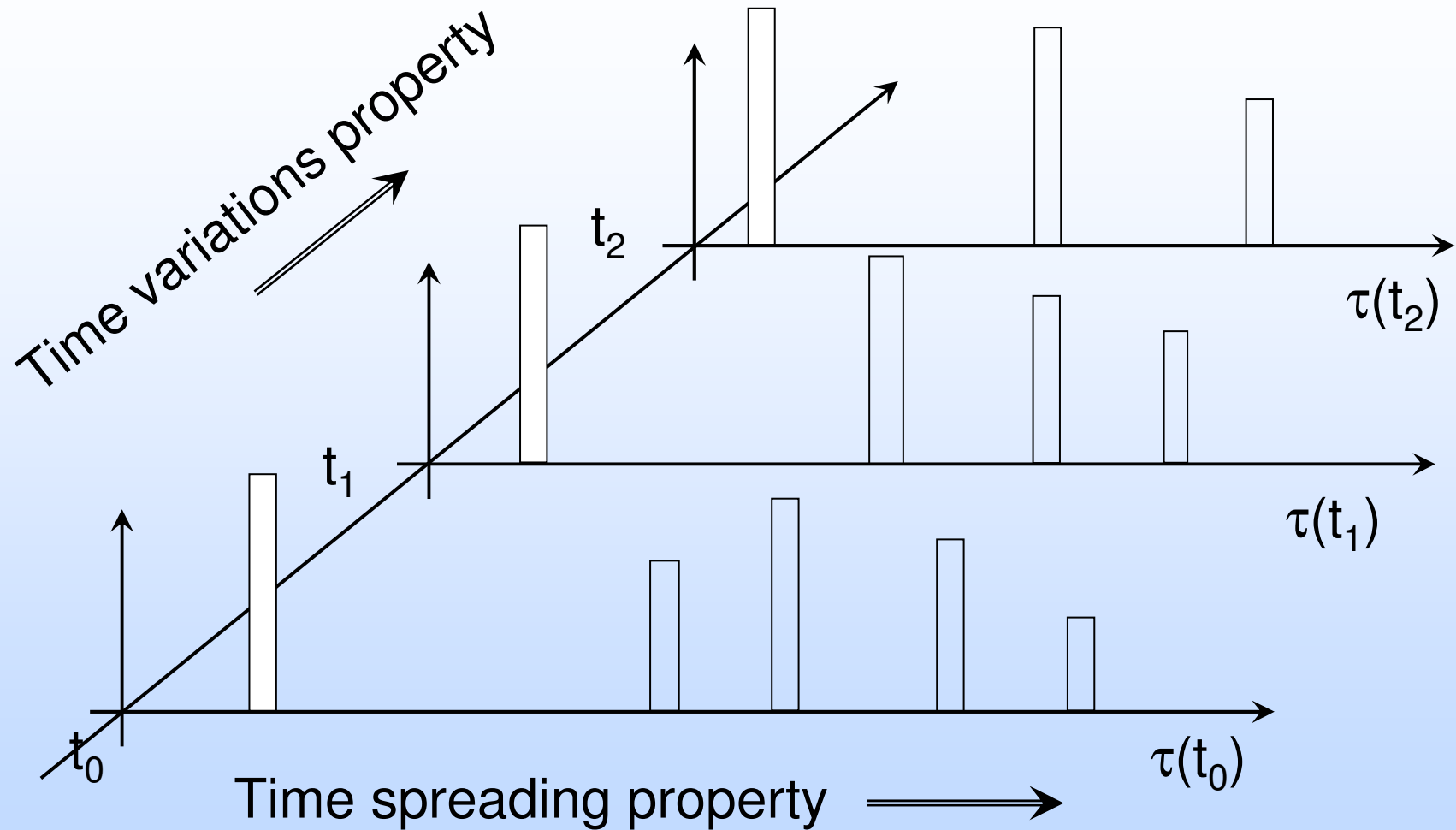
Matlab Assignment 1

- **Develop a WGN channel that accepts symbol input, adds noise to it with certain SNR and produces the noisy output.**
- **Develop a Flat Fading Channel following Rayleigh and rician distribution**
- **Develop a Frequency selective channel following Rayleigh and rician distribution**
 - All inputs to the channel are baseband signals.
 - Compare with the matlab functions (if exist)
 - Plot the probability of error vs SNR

Lecture one Fading channels

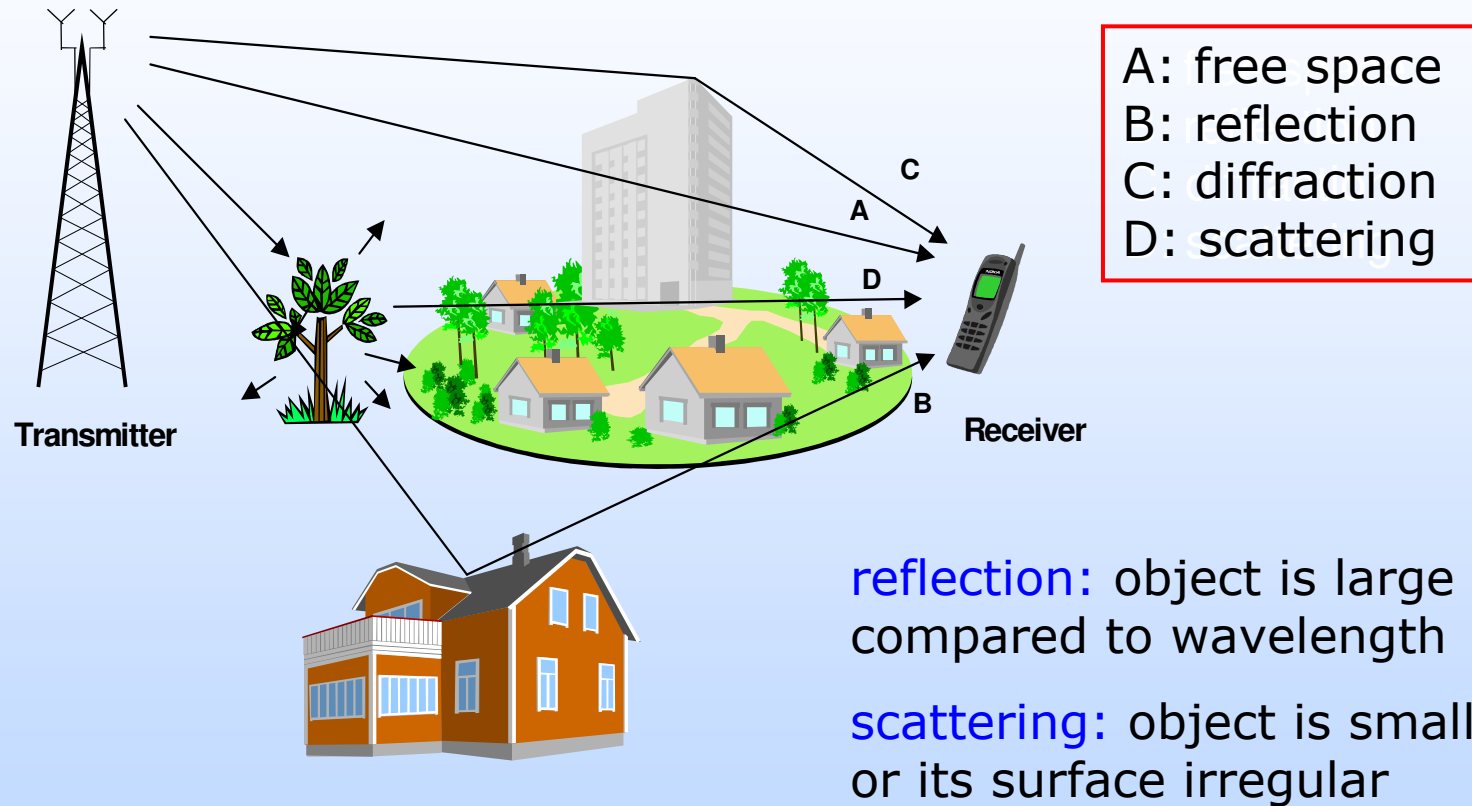


Impulse Response Characterization

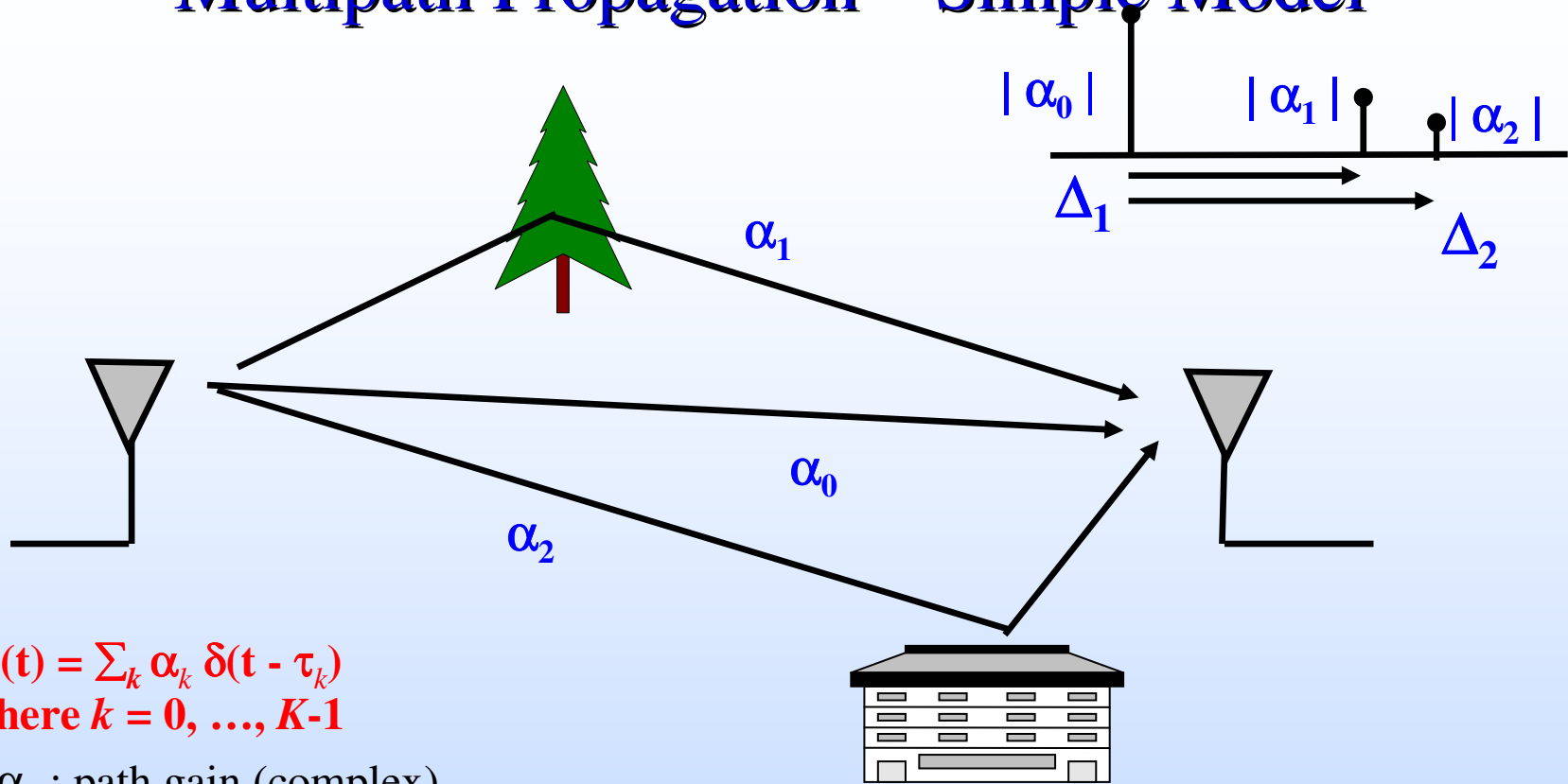


- Impulse response: Time-spreading : multipath
and time-variations: time-varying environment

Propagation mechanisms



Multipath Propagation – Simple Model



- $h_c(t) = \sum_k \alpha_k \delta(t - \tau_k)$
where $k = 0, \dots, K-1$

α_k : path gain (complex)

$\tau_0 = 0$ normalize relative delay of first path

$\Delta_k = \tau_k - \tau_0$ difference in time-of-flight

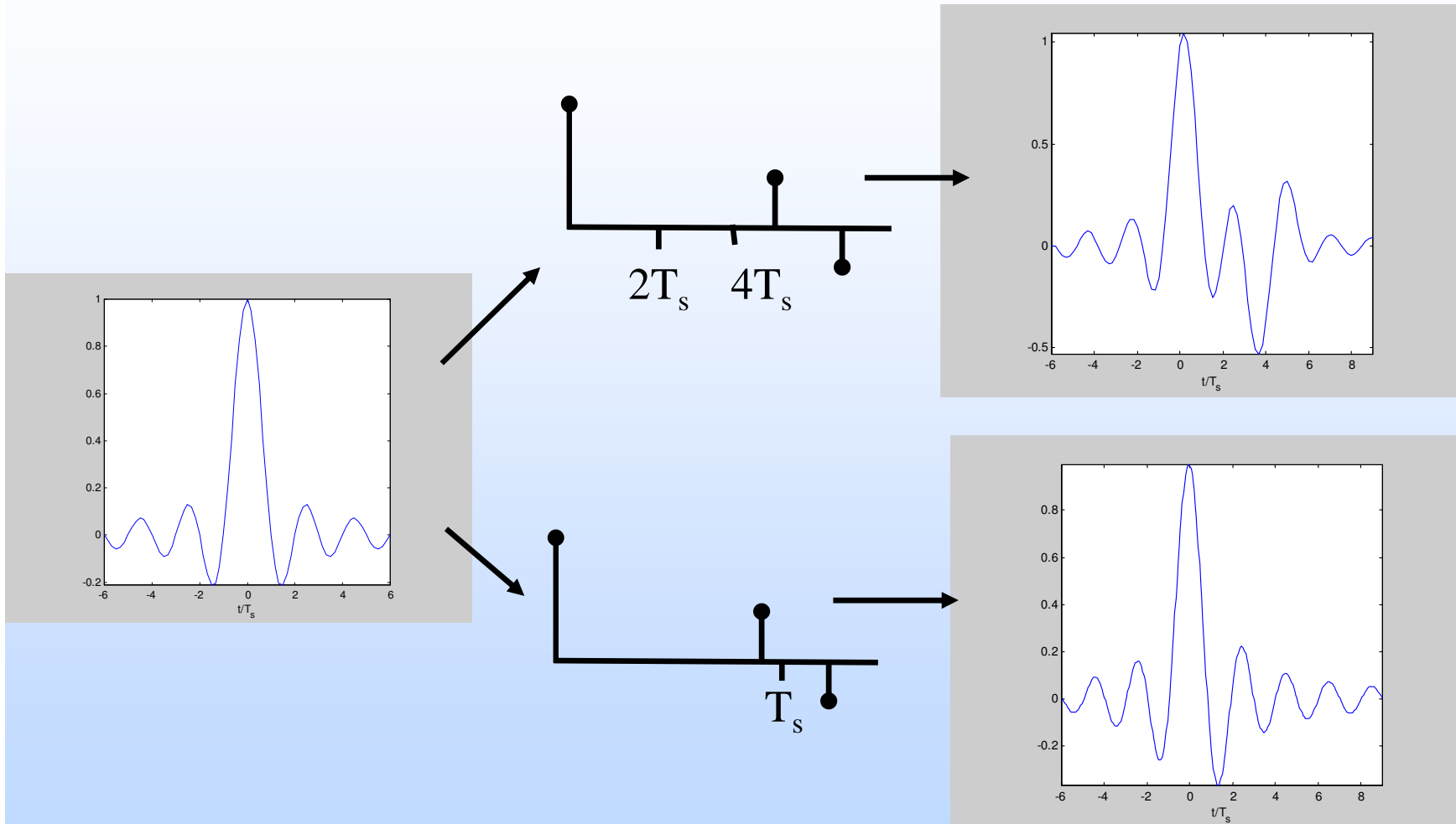
$$h(\tau, t) = \sum_{i=0}^{L-1} a_i(t) e^{j\phi_i(t)} \delta(\tau - \tau_i)$$

path attenuation

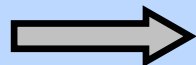
path phase

path delay

Impact of Multipath: Delay Spread & ISI



Max delay spread = effective number of symbol periods occupied by channel



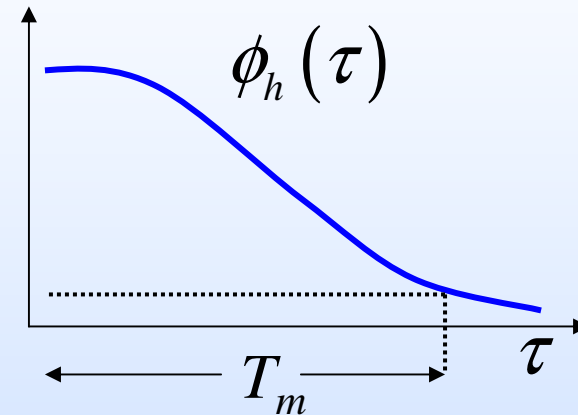
Requires equalization to remove resulting ISI

Stochastic (WSSUS) channel variables

Maximum delay spread: T_m

Maximum delay spread may be defined in several ways.

For this reason, the **RMS delay spread** is often used instead:

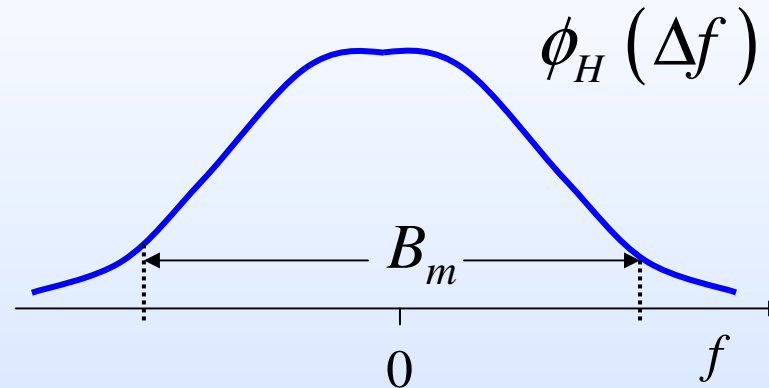


$$\sigma_\tau = \sqrt{\frac{\int \tau^2 \phi_h(\tau) d\tau}{\int \phi_h(\tau) d\tau} - \left[\frac{\int \tau \phi_h(\tau) d\tau}{\int \phi_h(\tau) d\tau} \right]^2}$$

Stochastic (WSSUS) channel variables

Coherence bandwidth
of channel:

$$B_m \approx 1/T_m$$



Implication of
coherence bandwidth:

If two sinusoids (frequencies) are spaced much less apart than B_m , their fading performance is similar.

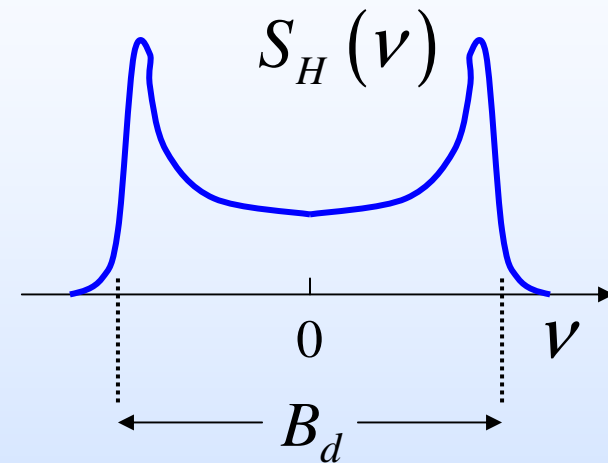
If the frequency separation is much larger than B_m , their fading performance is different.

Stochastical (WSSUS) channel variables

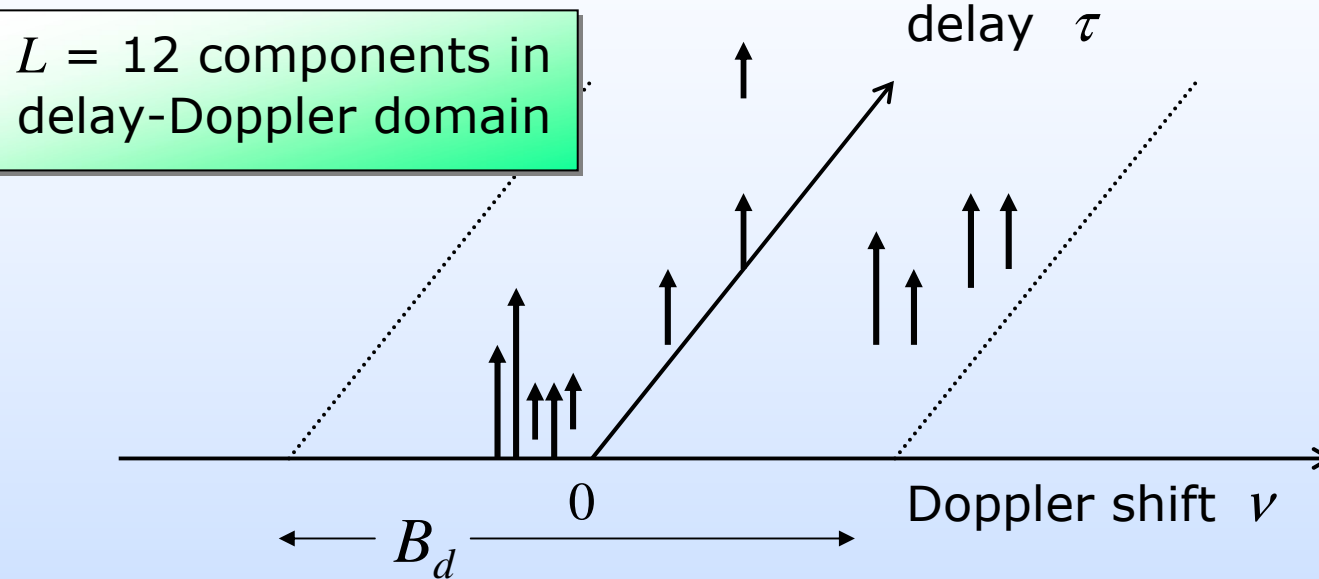
Maximum Doppler spread: B_d

The Doppler spectrum is often U-shaped (like in the figure on the right). The reason for this behaviour is the relationship

$$\nu = \frac{V}{\lambda} \cos \alpha = f_d \cos \alpha$$



Delay - Doppler spread of channel



$$h(\tau, t) = \sum_{i=0}^{L-1} a_i(t) e^{j(2\pi\nu_i t + \phi_i)} \delta(\tau - \tau_i)$$

Statistical Models

- **Design and performance analysis based on statistical ensemble of channels rather than specific physical channel.**
- **Rayleigh flat fading model: many small scattered paths**

$$h_\ell[m] \approx \sum_i a_i e^{-j2\pi f_c \tau_i}$$

Complex circular symmetric Gaussian .

$$h[m] \sim \mathcal{N}(0, \frac{1}{2}) + j\mathcal{N}(0, \frac{1}{2}) \sim \mathcal{CN}(0, 1)$$

- **Rician model: 1 line-of-sight plus scattered paths**

$$h[m] \sim \sqrt{\kappa} + \mathcal{CN}(0, 1)$$

Fading distributions (Rayleigh)

In a flat fading channel, the (time-variant) CIR reduces to a (time-variant) complex channel coefficient:

$$c(t) = a(t) e^{j\phi(t)} = x(t) + j y(t) = \sum_i a_i(t) e^{j\phi_i(t)}$$

When the quadrature components of the channel coefficient are **independently and Gaussian distributed**, we get:

$$p(a) = \frac{a}{\sigma^2} e^{-a^2/2\sigma^2}$$

Rayleigh distribution

$$p(\phi) = \frac{1}{2\pi}$$

Uniform distribution

Fading distributions (Rice)

In case there is a strong (e.g., LOS) multipath component in addition to the complex Gaussian component, we obtain:

$$c(t) = a_0 + a(t)e^{j\phi(t)} = a_0 + \sum_i a_i(t)e^{j\phi_i(t)}$$

From the joint (magnitude and phase) pdf we can derive:

$$p(a) = \frac{a}{\sigma^2} e^{-(a^2+a_0^2)/2\sigma^2} I_0\left(\frac{aa_0}{\sigma^2}\right)$$

Rice distribution

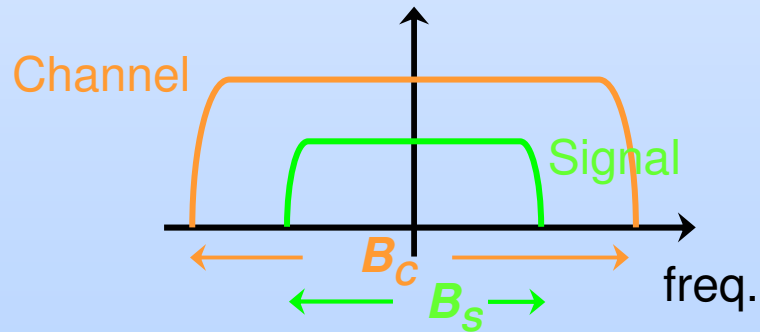
Modified Bessel function of first kind and order zero

Channel Classification

Based on Time-Spreading

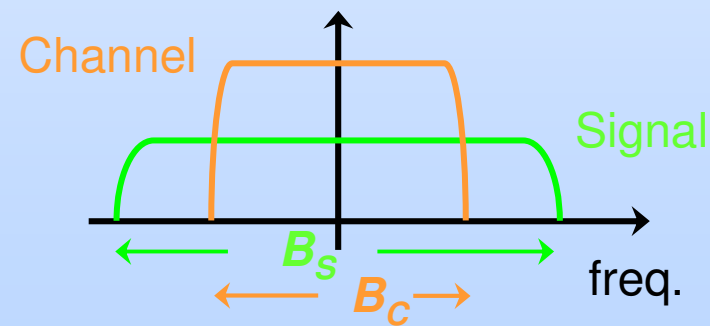
Flat Fading

1. $B_S < B_C \Leftrightarrow T_m < T_s$
2. Rayleigh, Ricean distrib.
3. Spectral char. of transmitted signal preserved



Frequency Selective

1. $B_S > B_C \Leftrightarrow T_m > T_s$
2. Intersymbol Interference
3. Spectral chara. of transmitted signal not preserved
4. Multipath components resolved

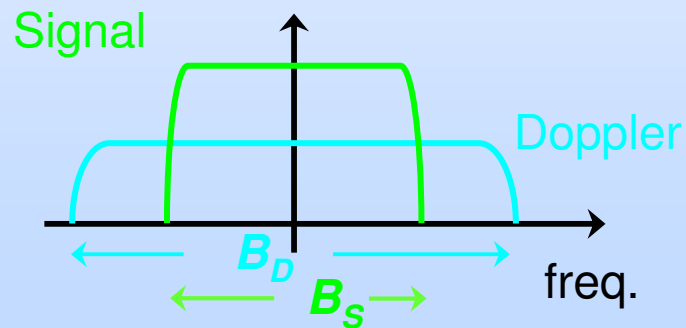


Channel Classification

Based on Time-Variations

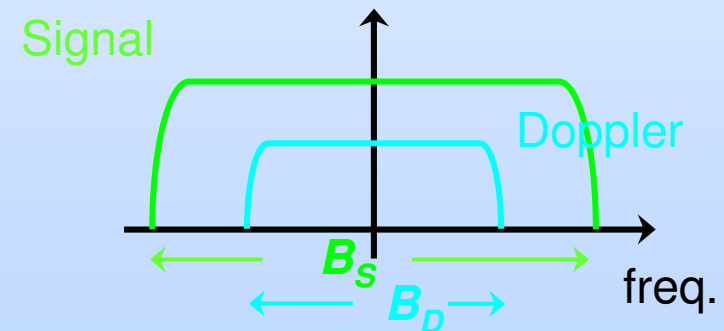
Fast Fading

1. High Doppler Spread
2. $1/B_d \cong T_C < T_s$

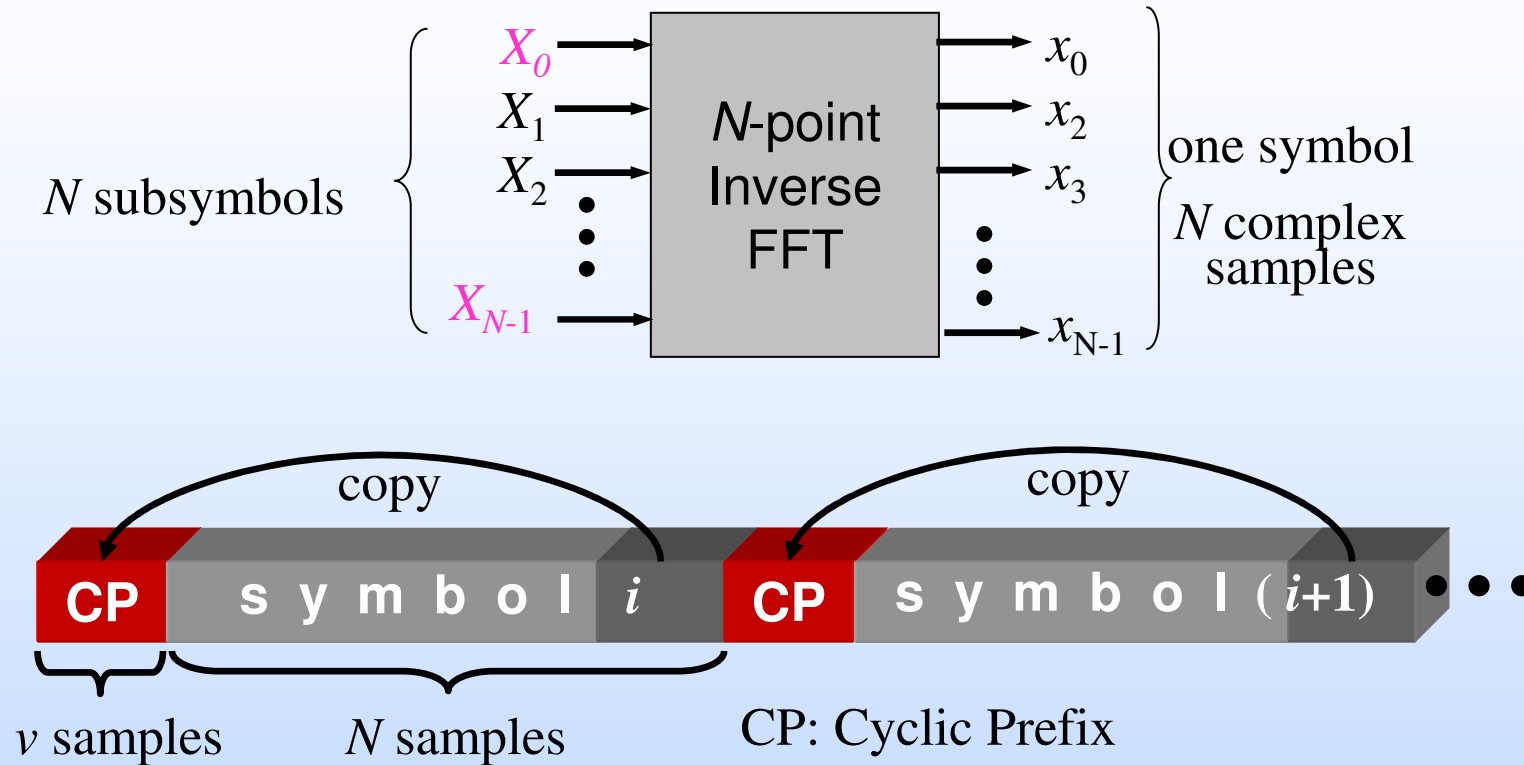


Slow Fading

1. Low Doppler Spread
2. $1/B_d \cong T_C > T_s$



An OFDM Symbol



- **Bandpass transmission allows for complex waveforms**

- **Transmit:** $y(t) = \text{Re}\{(I(t)+j Q(t)) \exp(j2\pi f_c t)\}$
 $= I(t) \cos(2\pi f_c t) - Q(t) \sin(2 \pi f_c t)$

Introduction to OFDM

- Basic idea
 - » Using a large number of parallel narrow-band sub-carriers instead of a single wide-band carrier to transport information
- Advantages
 - » Very easy and efficient in dealing with multi-path
 - » Robust against narrow-band interference
- Disadvantages
 - » Sensitive to frequency offset and phase noise
 - » Peak-to-average problem reduces the power efficiency of RF amplifier at the transmitter
- Adopted for various standards
 - DSL, 802.11a, DAB, DVB

OFDM can achieve large delay spread tolerance at high bit rates by:

Converting single bit stream in N parallel bit streams

- Symbol duration is increased, so relative delay spread decreases
- Each parallel bit stream is modulated on one of N subcarriers

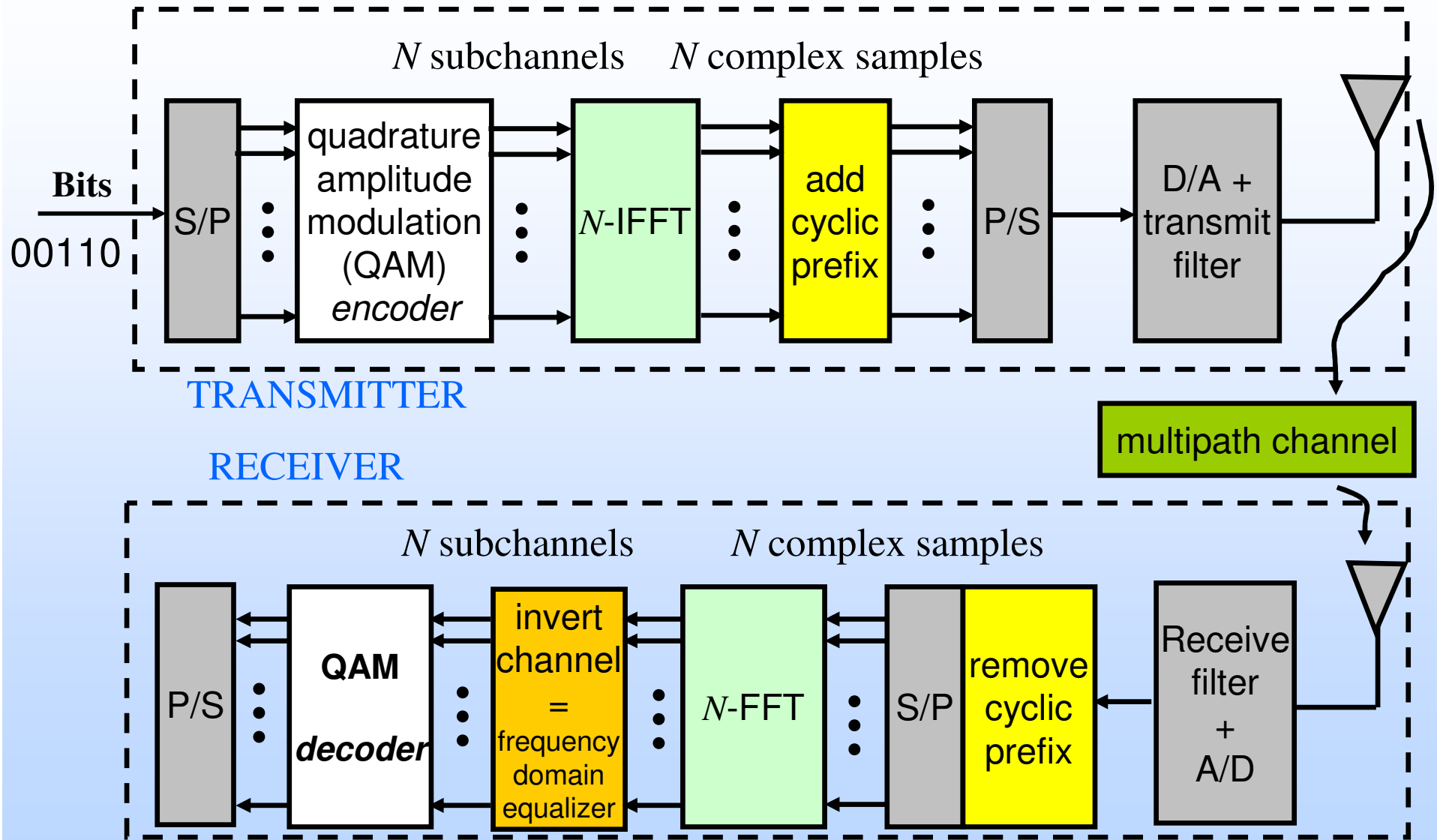
Adding a guard time to each OFDM symbol

- Inter Symbol Interference (ISI) is avoided
- Guard loss is made small (<1 dB) by choosing N large enough

Use FEC coding to correct for subcarriers in deep fades

- In a multipath channel, subcarriers have different amplitudes
- By using coding, performance is determined by average received power rather than lowest subcarrier power

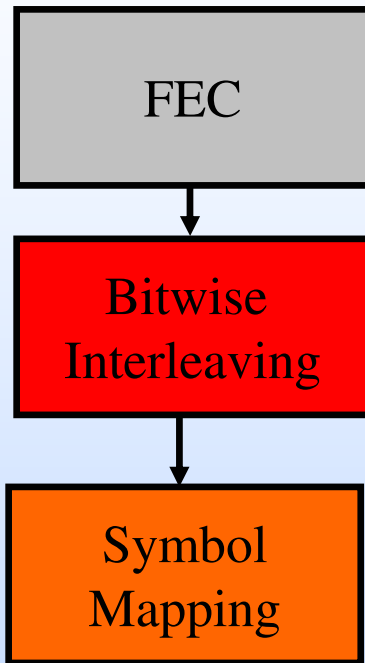
An OFDM Modem



Coded OFDM (COFDM)

- **Error correction is *necessary* in OFDM systems**
- **Forward error correction (FEC)**
 - Adds redundancy to data stream
 - Examples: convolutional codes, block codes
 - Mitigates the effects of bad channels
 - Reduces overall throughput according to the coding rate k/n
- **Automatic repeat request (ARQ)**
 - Adds error detecting ability to data stream
 - Examples: 16-bit cyclic redundancy code
 - Used to detect errors in an OFDM symbol
 - Bad packets are retransmitted (hopefully the channel changes)
 - Usually used with FEC
 - Minus: Ineffective in broadcast systems

Typical Coded OFDM Encoder



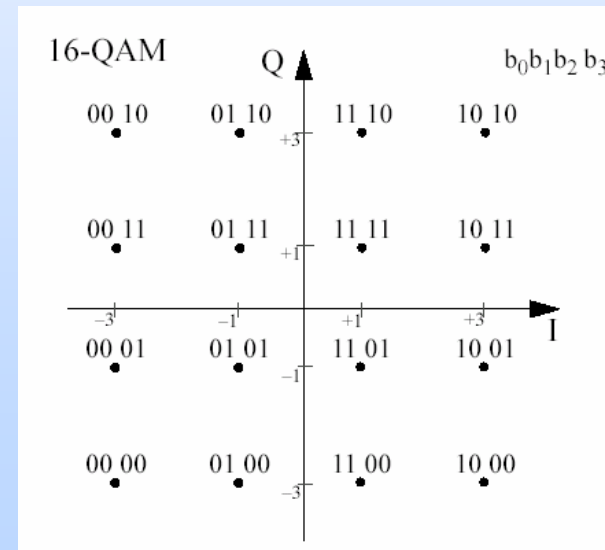
- Reed-Solomon and/or convolutional code



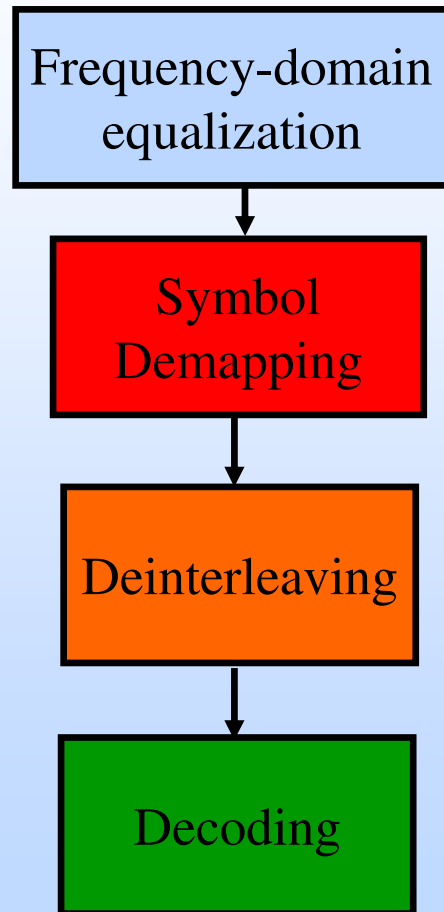
Rate 1/2

- Intersperse coded and uncoded bits

- Map bits to symbols



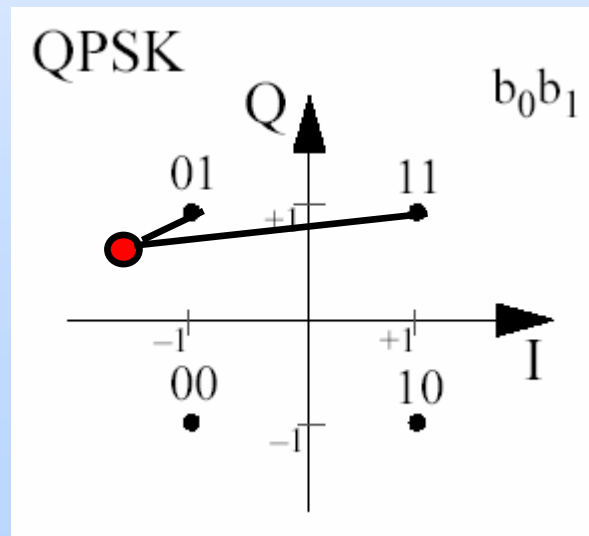
Typical Coded OFDM Decoder



- **Symbol demapping**

- Produce soft estimate of each bit
- Improves decoding

$$L(b_k) = \log \left(\frac{Pr(x|b_k = 1)}{Pr(x|b_k = 0)} \right)$$
$$\approx \log \left(\frac{\max_{s_k} Pr(x|b_k = 1, s_k)}{\max_{s_k} Pr(x|b_k = 0, s_k)} \right)$$



OFDM Mathematics

$$s(t) = \sum_{k=0}^{N-1} X_k e^{j2\pi f_k t} \quad t \equiv [0, T_{os}]$$

Orthogonality Condition

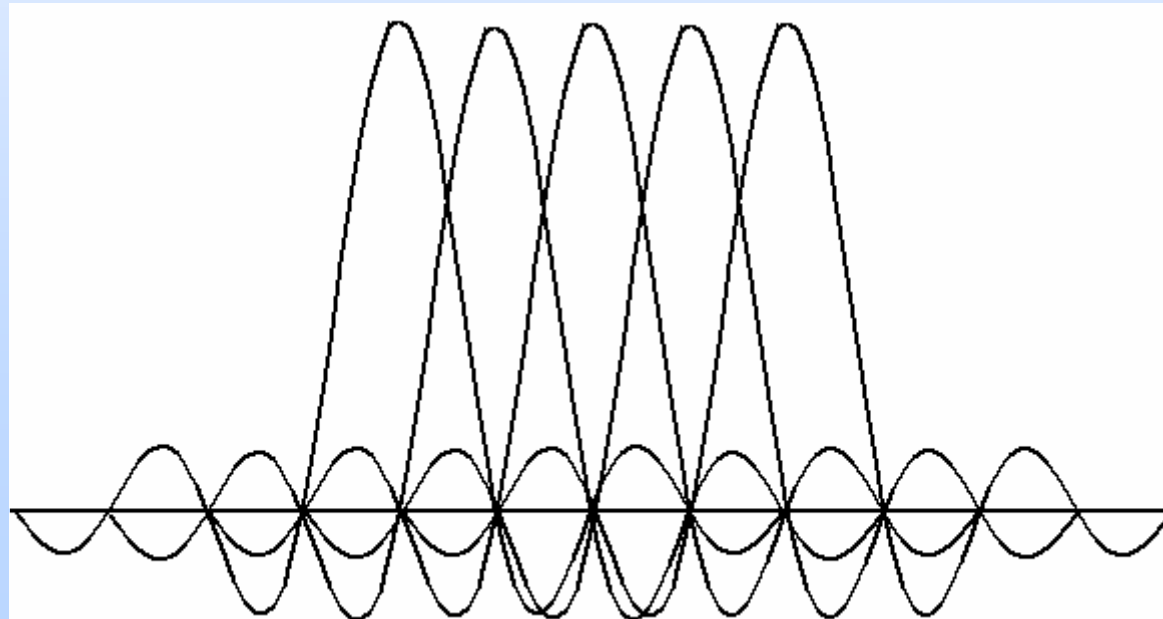
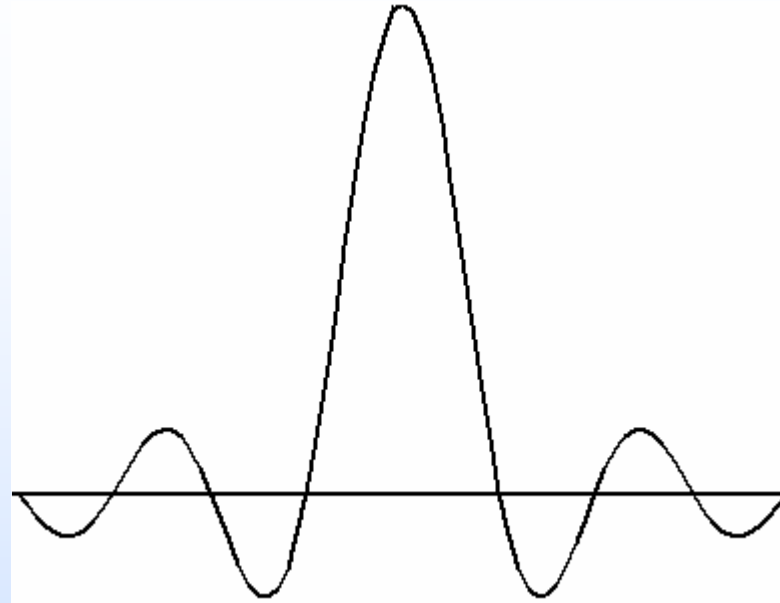
$$\int_0^{T_{os}} g_1(t) \cdot g_2^*(t) dt = 0$$

In our case

$$\int_0^{T_{os}} e^{j2\pi f_p t} \cdot e^{-j2\pi f_q t} dt = 0$$

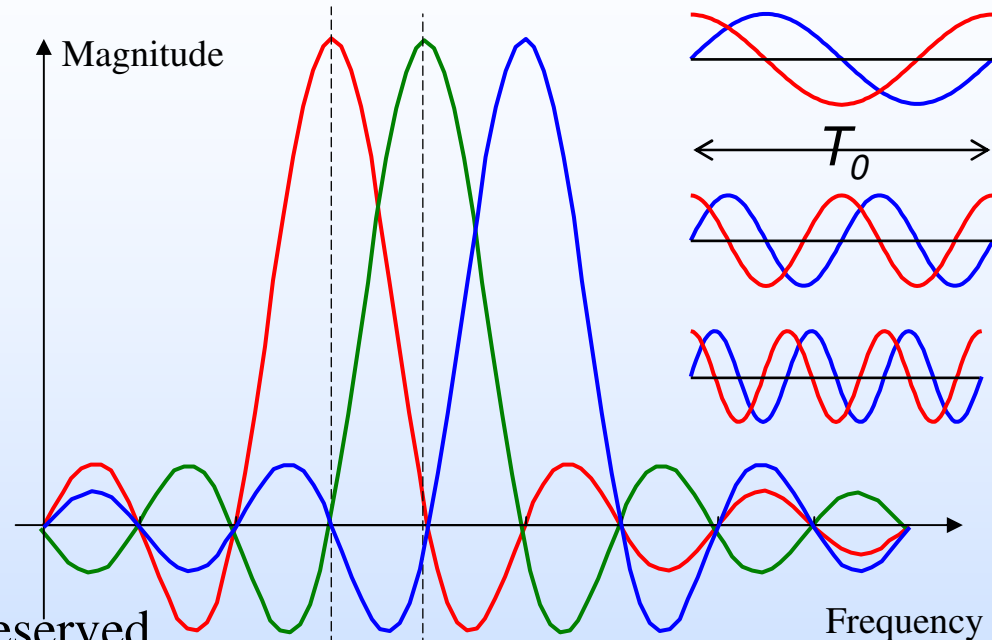
For $p \neq q$ Where $f_k = k/T_{os}$

Transmitted Spectrum



Spectrum of the modulated data symbols

- **Rectangular Window of duration T_0**
- **Has a sinc-spectrum with zeros at $1/T_0$**
- **Other carriers are put in these zeros**
- **→ sub-carriers are orthogonal**

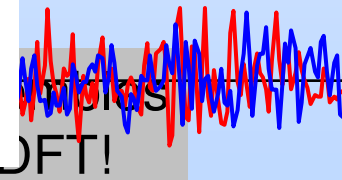


Subcarrier orthogonality must be preserved

Compromised by timing jitter, frequency offset, and fading.

N sub-carriers:

$$s_{BB,k}(t) = w(t - kT) \sum_{i=0}^{N-1} x_{i,k} e^{j2\pi i \Delta f (t - kT)}$$



OFDM terminology

- Orthogonal carriers referred to as subcarriers $\{f_i, i=0, \dots, N-1\}$.
- OFDM symbol period $\{T_{os} = N \times T_s\}$.
- Subcarrier spacing $\Delta f = 1/T_{os}$.

OFDM and FFT

- Samples of the multicarrier signal can be obtained using the IFFT of the data symbols - a key issue.
- FFT can be used at the receiver to obtain the data symbols.
- No need for 'N' oscillators, filters etc.
- Popularity of OFDM is due to the use of IFFT/FFT which have efficient implementations.

OFDM Signal

$$s(t) = \sum_{n=-\infty}^{\infty} \left(\sum_{k=0}^{N-1} X_{n,k} g_k(t - nT_{os}) \right)$$

$$g_k(t) = \begin{cases} e^{j 2 \pi f_k t} & t \equiv [0, T_{os}] \\ 0 & \text{Otherwise} \end{cases}$$

$$f_k = \frac{k}{T_{os}} \quad K=0, \dots, N-1$$

By sampling the low pass equivalent signal at a rate N times higher than the OFDM symbol rate $1/T_{os}$, OFDM frame can be expressed as:

$$F_n(m) = \sum_{k=0}^{N-1} X_{n,k} g_k(t - nT_{os}) \Big|_{t = (n + \frac{m}{N})T_{os}} \quad m = 0 \dots N-1$$

$$F_n(m) = \left(\sum_{k=0}^{N-1} X_{n,k} e^{j2\pi k \frac{m}{N}} \right) = N.IDFT\{X_{n,k}\}$$

Interpretation of IFFT&FFT

- **IFFT at the transmitter & FFT at the receiver**
- **Data symbols modulate the spectrum and the time domain symbols are obtained using the IFFT.**
- **Time domain symbols are then sent on the channel.**
- **FFT at the receiver to obtain the data.**

Assume the information data sequence be $X(k)$, $k=0, 1 \dots N-1$, then the transmitted signal $s(t)$ can be expressed as

$$s(t) = \sum_{p=0}^{N-1} X(p) \exp(j2\pi f_p t) \\ = \sum_{p=0}^{N-1} X(p) \exp(j2\pi p \Delta f t),$$

valid for $0 \leq t < T$; $0 \leq p \leq N-1$.

The received signal $r(t)$ is

$$r(t) = s(t) * h(t) + w(t)$$

The received on subcarrier k_0 is

$$Y(k) = \frac{1}{T} \int_0^T r(t) \exp(-j2\pi k \Delta f t) dt$$

Assume $h(t)=1$, $w(t)=0$ (ideal channel), then $Y(k)$ becomes

$$Y(k) = \frac{1}{T} \int_0^T \sum_{p=0}^{N-1} X(p) \exp(j2\pi p \Delta f t) \exp(-j2\pi k \Delta f t) dt$$

Transmitted data can be fully recovered from above equation, if $P = k$ in $Y(K)$.

